

# Design of the opto-electronic receiver for deep space optical communications

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## ABSTRACT

The opto-electronic receiver (detector and pre-amplifier) necessary to meet the demands of high capacity deep space missions is designed for a Mars-Earth optical communication link. The receiver requirements are driven by link performance (data rate, bit-error rate, margin), delivered power, pulse width, background signal, telescope quality, and atmospheric effects. Meeting these requirements becomes more challenging as the mission range and the demand for link capacity increases. In this article, the detector's characteristics (e.g. quantum efficiency, noise, gain, and diameter) are designed to address these various requirements. The receiver sensitivity's dependence on the background noise power and on the APD detector's characteristics is analyzed. The improvement in opto-electronic receiver sensitivity is quantified for improvements in APD quantum efficiency, ionization factor, and bulk dark current. It is also found that as the background signal increases the improvement on the receiver sensitivity from an improved detector is diminished due to the quantum noise limit. An opto-electronic receiver is designed based on a Silicon APD to meet the mission requirement of a PPM (M=256) 30 kbps data rate (BER of  $10^{-5}$ ) link. Improvements to the APD detector are also studied to describe a design that would achieve over 50 kbps data rates for a Mars-Earth optical communication link.

**Keywords:** Opto-electronic receiver, free-space communications, deep space optical communications, background noise power, APD.

## 1. INTRODUCTION

JPL/NASA plans to develop a fully functional Deep Space Optical Receiver (DSOR) to support optical communication needs<sup>1</sup>. The DSOR detects a pulse-position-modulated (PPM) signal from the spacecraft with an avalanche photodiode (APD) detector, whose signal is then amplified and processed to extract the data. The front-end detector is the critical component of this direct detection optical communication deep space receiver. The detector must maintain low-noise as it converts the signal photons into an electrical signal that can be post-processed to obtain the communication signal. The figure of merit used for this conversion (a.k.a. receiver sensitivity) is determined by the detector's quantum efficiency, intrinsic gain and noise characteristics. Furthermore, in order to deliver a receiver capable of making a free-space link with high data rates and low bit-error rate (BER), the detector must minimize background noise and atmospheric effects by controlling its collection area.

The opto-electronic receiver detector and pre-amplifier specifications will be derived from analysis of the dependence of the receiver's performance on background noise and the detector's characteristics and their impact on the link design constrained by mission link performance goals. Also, analysis of the estimated amount of background light that would be collected during daytime operation by an optical telescope on the earth's surface supporting an optical downlink from a Mars orbiter will be presented. Estimates of the blur circle for the different telescope's being considered will also be presented. Simulation results on the receiver sensitivity for state-of the art detectors will also be presented.

The opto-electronic receiver is designed for a reference Mars mission. The transmitter baselined for implementing such a communication link is a Q-switched Nd:YAG laser<sup>2</sup>. The specifications of the transmitter are tabulated in Table 1.

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From the receiver's point-of-view a certain level of the transmitted photons (received power) must be collected to detect the pulse. This level is determined from the values in Table 1 and the particular communication link design goals/requirements. This is usually stated in a design control table for the communication link. In the design of the opto-electronic receiver for this article, the values summarized in Table 2 will be used.

Parameter	Specification
Laser Type	Q-switched Nd:YAG
Output power	1 Watt
Downlink Wavelength	1064 nm
Pulse Width	25 ns
Modulation Extinction Ratio	$10^{-5}$
Telescope Diameter	10 cm
Transmitter Optics Losses	28 %
Modulation Format	PPM (M=256)
Coding	Reed-Solomon

Table 1 Spacecraft Transmitter Specification.

Parameter	Specification
Required link margin	3 dB
Data Rate	30 kbps
Bit Error Rate	$10^{-5}$
Link Range	2 AU (300 million km)
Atmospheric Losses	32.8 %
Receiver telescope optics losses	55.1 %

Table 2 Link design parameters.

## 2. OPTO-ELECTRONIC DEEP SPACE RECEIVER DESIGN

The major factors to consider in the design of the opto-electronic receiver are size (diameter), speed and sensitivity. The size of the detector must be large enough to efficiently collect the photons but not so large that it increases noise or reduces bandwidth. The major design drivers on the size of the detector are the blur circle and the bandwidth requirements. The speed of the receiver is dominated by the capacitance of the detector. The major driver for the speed is the desired data rate and the pulse width of the received signal. The sensitivity of the opto-electronic receiver must be sufficient to provide the required link margin. Many factors influence the sensitivity, for example, detector quantum efficiency, gain, and noise characteristics. The drivers for a sensitive design are also numerous, such as, link design, background signal, BER, wavelength and detector characteristics. All these factors will be considered in the design of the opto-electronic receiver in this section.

### 2.1 Size

The diameter of the detector needs to be specified such that it collects all the focused energy that has been received by the optical telescope. But, this goal is hampered by the fact that the focus of the spot is not diffraction limited due to the influence of atmospheric turbulence. This extended spot, for a desired percentage of encircled energy, is also known as the 'blur circle.' In appendix A, the dependence of the focus spot size on atmospheric turbulence is given.

Figure 1 is a contour plot of the fraction  $P(r)$  of the incident energy collected within the normalized detector diameter ( $2a$ ) as a function of normalized radius  $r (=a/\lambda F)$  and normalized telescope diameter  $D/r_0$ . Consider a telescope having a circular entrance pupil of diameter  $D$  and a system focal ratio  $F$ . A typical nighttime value of atmospheric coherence length  $r_0$  is 10 to 20 cm, corresponding to an atmospheric seeing disk radius  $\lambda/r_0 \sim 5$ -10  $\mu\text{rad} \sim 1$ -2 arc-seconds at  $\lambda = 1.06 \mu\text{m}$ . Typical daytime values for  $r_0$  are 5 to 10 times smaller<sup>3</sup>. Unfortunately, the statistical dependence of  $r_0$  on solar zenith angle, observation zenith angle, and season is not well documented in the literature, especially for solar zenith angles corresponding to daytime operations. For the purpose of discussion, then, it is assumed that coherence cell sizes of interest range from 20 cm ('nighttime') to 2 cm ('daytime').

The detector diameter necessary can be obtained from Figure 1 such that it encircles 80 % of the energy. For example, Figure 1 shows that if the telescope diameter is 3.5 meters and is operated under daytime conditions ( $r_0 = 2$  cm), a normalized detector radius of  $r = 160$  will be required to encircle 80 % of the signal. If the system focal ratio is  $F = 2$ , the corresponding detector diameter would be  $2a = 680 \mu\text{m}$ . Table 3 lists these diameters for three telescope apertures. The actual size of the telescope can then be selected to meet the link objectives. Using the ‘daytime’ condition to obtain the required detector diameter yields the benefit that the percentage of encircled energy in less severe turbulence will be higher. For example, during ‘nighttime’ operation the typical coherence cell size is 20 cm, which would cause over 98 % of the energy to fall within the diameter of the same detector. Of course, to encircle a larger percentage of the energy during daytime conditions the required diameter would increase.

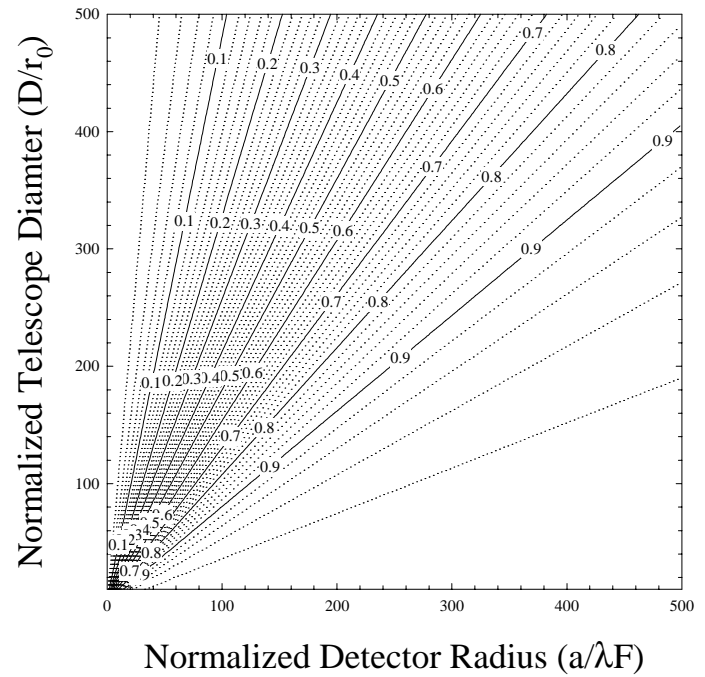


Figure 1. Contours show the fraction of the incident energy encircled as a function of the normalized detector radius and the differing ratios of telescope diameter to atmospheric coherence cell size.

Telescope Aperture Diameter (m)	Detector Radius, normalized ( $a\lambda/F$ )	Required Detector Diameter (mm)
1	50	0.20
3.5	160	0.68
10	460	2.0

Table 3. Detector diameter for various telescope aperture sizes. Assuming wavelength of 1064 nm and focal ratio  $F = 2$  with 80 % encircled energy and an atmospheric coherence cell size of 2 cm (‘daytime’).

## 2.2 Speed and detector selection

As seen in section 2.1, the detector’s diameter must be very large to mitigate the effects of atmospheric turbulence. The speed requirements for the detector are based on the 25 ns downlink pulse-width. Therefore, a detector with a 2.5 nsec rise/fall time (90-10 %) is sufficient to respond to the incoming pulse modulations. In practice this a relatively moderate speed requirement. The sensitivity of the detector selected needs to be highest possible. The device parameters that impact high sensitivity are high quantum efficiency (QE at 1064 nm for this communication link), low noise characteristics and sufficient gain to amplify the signal. The sensitivity required for the link in this article will be derived in section 2.3.

The photo-multiplying tube(PMT), the optical pre-amplifier (OA) and the avalanche photodetector (APD) were considered as the front-end component in the design of this receiver. The photo-multiplying tube was found to have extremely high gain, but low quantum efficiencies at 1064 nm. Optical pre-amplifiers were found to provide high gain, but had small

collection areas. The APD was selected as the front-end component in this receiver design based on the availability of devices with high QE's, low-noise and sufficient gain at our wavelength of interest.

## 2.3 Sensitivity

The sensitivity of the opto-electronic receiver is dependent on a large number of parameters. The major factors are background noise power, detector characteristics, modulation format and pre-amplifier equivalent noise density. These in turn have their sub-dependencies. The background signal level is affected by the telescope receiver aperture diameter, the receiver optical efficiency, any optical filtering, the detector field-of-view, any light sources in the field-of-view, and by the background irradiance (or spectral radiance). The APD detector dependencies are QE, gain, k-factor, bulk current and surface current. The modulation format parameters have been set by the choice of link design outlined in section 1. These format details include the PPM format, the modulation slot width to dead time ratio, the modulation extinction ratio and the bit error rate. In this section, the required sensitivity of the receiver to meet the link requirements will be obtained. First a bound for the estimated background noise power will be derived, then the opto-electronic receiver sensitivity dependence on the dominant terms of background signal and detector characteristics will be explored. This section will conclude by obtaining the receiver design necessary that meet the link objectives.

### 2.3.1 Background noise power

Background noise power, broadly defined, is any non-signal photon exciting the detector of an optical system. Previous work<sup>4,5</sup> has enumerated the various sources of background light and shown that diffuse sky brightness and sunlight scattered from the surface of Mars (Marslight) will be its chief constituents for a Mars optical downlink. For this reason the discussion of background light will be restricted to these two sources. In appendix A, the amount of background light is estimated for these two sources in order to bound the estimated level of noise power incident on the detector.

The typical diffuse sky radiance  $E$  is approximately<sup>5</sup> 10 mW/m<sup>2</sup>/sr/nm at 1.06  $\mu\text{m}$  for clear skies. Measured sky radiance at  $\lambda = 0.667 \mu\text{m}$  varies by at least factor of four with the portion of sky examined<sup>8</sup>, as sky radiance increases sharply close to the sun<sup>3</sup>. The variation will likely be greater at longer wavelengths that are less likely to scatter and therefore provide a less-isotropic diffuse sky radiance. For this reason, the sky brightness curve in Figure 2 should be interpreted as a range of sky brightness that depends strongly on the minimum sun-earth-probe angle. The flux entering the normalized detector is independent of the telescope diameter. Although a larger telescope captures more diffuse skylight, the increase is compensated by the reduction in FOV for a fixed focal ratio. Table 4 summarizes the typical number of diffuse sky background photons for the three detector choices being considered in the design.

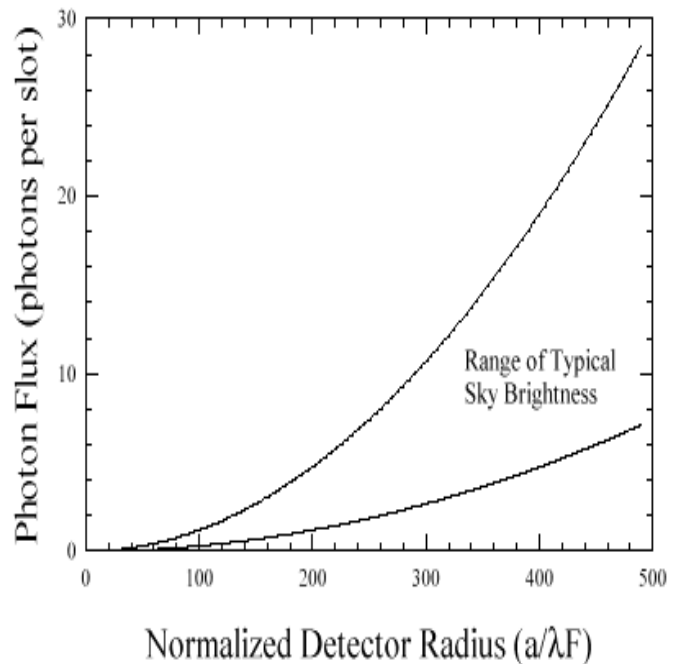


Figure 2. The typical number of diffuse sky background photons collected versus the normalized detector size.

Detector Radius, normalized ( $a\lambda/F$ )	Diffuse Sky Background (photons/ns)
50	0.1
160	1 - 3
460	6 - 25

Table 4. Diffuse sky background for various detector sizes. The optical filter bandwidth is 0.2 nm and the wavelength is at 1060 nm.

Figure 3 shows the number of Mars background counts per nanosecond as a function of telescope diameter and normalized detector radius. For  $r = 160$  (corresponding to a 3.5 meter diameter telescope collecting 80 % of the signal energy during daytime conditions), less than 1 photon per nanosecond will be found. It is evident in Figure 3 that for a fixed telescope diameter the background from Marslight ceases to increase beyond the normalized detector diameter at which the detector's field-of-view encompasses Mars entirely. There is a corresponding effect for a fixed normalized detector diameter as a function of increasing telescope diameter. Although the detector's normalized diameter is fixed, its field-of-view decreases with increasing telescope aperture. The decrease in detector field-of-view exactly compensates the increased telescope collection aperture and the collected Marslight becomes independent of telescope diameter. In both cases, it is assumed that the seeing limit is smaller than angular diameter of Mars at favorable opposition so that a reasonably uniformly filled image is formed. The background signal level of Marslight is tabulated in Table 5 for the three telescope diameters under consideration. Because the Marslight calculations are done when Mars is at favorable opposition, the value estimated for Marslight background represents the upper bound of signal from the planet.

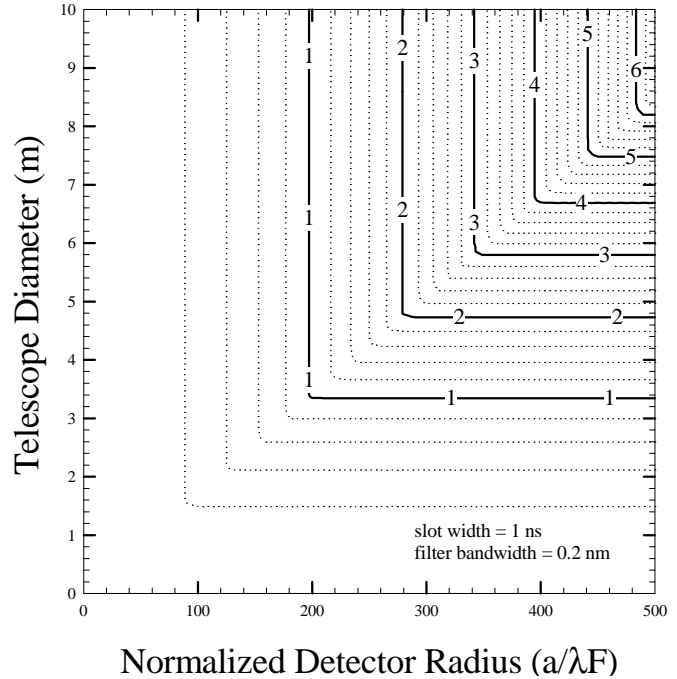


Figure 3. Contours show the number of Mars background photons/ns for different telescope diameters and detector sizes.

Telescope Aperture Diameter (m)	Detector Radius, normalized ( $a\lambda/F$ )	Marslight Background (photons/ns)
1	50	0.1
3.5	160	0.6
10	460	5

Table 5. Marslight signal for various detectors. Optical filter bandwidth is 0.2 nm and wavelength is 1060 nm.

The plots presented in this section scale linearly for different slot durations and spectral filter passbands. For example, if the slot duration was increased to 31.25 ns, corresponding to the receiver considered for this link, the typical collected diffuse skylight and Marslight would increase 31.25 times to 31.25 photons per slot and 18.8 photons per slot, respectively, for a 3.5-meter telescope. If the spectral filter pass-band was 2 nm instead of 0.2 nm, the typical collected diffuse skylight and Marslight would increase another 10 times to 313 photons per slot and 188 photons per slot, respectively. The two sources

of background signal can be summed to obtain a bound on the estimated background signal, see Table 6, which would be incident on the detector. The lowest level is obtained during 'nighttime' conditions, which is when the diffused sky level is negligible (typically 3-5 orders of magnitude smaller from its daytime value). In this case the background signal is dominated by the planet (Mars) being in the field of view. For 'daytime' conditions, the level is the sum of the planet light and the diffused sky. The two conditions are tabulated to demonstrate the dependence on telescope diameter and detector radius. These values represent the lower and upper values estimated for the background noise power for a Mars link.

Telescope Aperture Diameter (m)	Detector Radius, Normalized ( $a\lambda/F$ )	Total Night Background (photons/slot)	Total Day Background (photons/slot)
1	50	3	6 – 24
3.5	160	19	31 – 94
10	460	156	344 – 938

Table 6. Total background signal for 'nighttime' and 'daytime' conditions for various detectors/telescopes. Optical filter bandwidth is 0.2 nm, wavelength is 1060 nm and receiver slot width is 31.25 ns.

### 2.3.2 Sensitivity dependence on background signal

The existence of background noise power on the detector has the effect of reducing the sensitivity of the receiver system. This is demonstrated in Figure 4, where the 'receiver required power' to make the link with our APD-based receiver is plotted. This receiver sensitivity calculation is done using the direct detection link parameters from section 1 (eg. BER =  $10^{-5}$ , slot width = 31.25 ns,  $\lambda$  = 1064 nm) and with Silicon APD parameters (QE of 0.38,  $k$  = 0.02, max. gain of 100, surface current = 100 nA, bulk current = 10 pA). The pre-amplifier used in this design is a high-impedance type with a noise equivalent current of 0.13 pA/ $\sqrt{\text{Hz}}$ . These parameters represent state-of-the-art devices available from commercial sources. The analysis was done using a model based on Gaussian statistics<sup>6</sup> for the APD and implemented in a spreadsheet for ease of use<sup>7</sup>.

Figure 4 shows the dependency of the receiver sensitivity as a function of background noise power. As expected, for high levels of background signal the sensitivity becomes quantum noise limited, while at low levels the performance is device noise limited. On this plot, the range of background signal levels estimated for the three telescope aperture diameters being considered is superimposed. It is clear that for larger telescope diameters the performance of the opto-electronic receiver tends to be quantum noise limited. The range indicated by the bar for each telescope size describes the amount of background signal in going from 'nighttime' to 'daytime' conditions, see Table 6 for values.

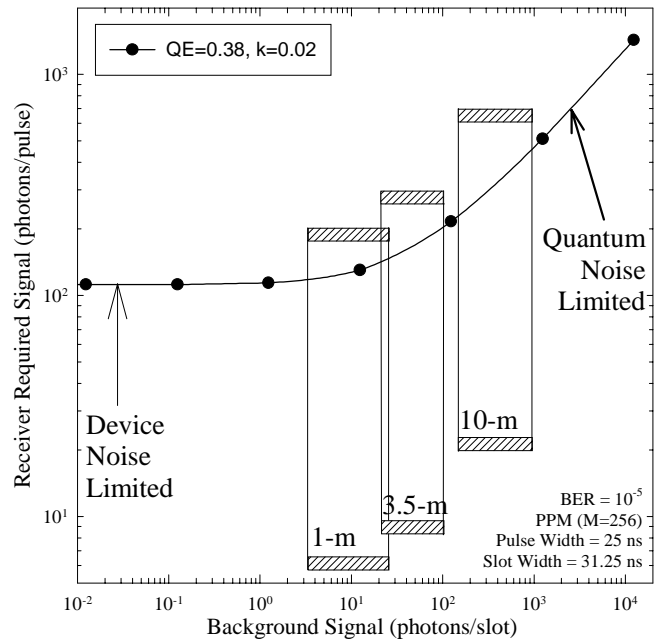


Figure 4. Receiver sensitivity dependence on background signal.

### 2.3.3 Sensitivity dependence on detector device parameters

As mentioned earlier, the sensitivity of the opto-electronic receiver depends strongly on the detector device characteristics. In this section, this dependence is demonstrated in order to determine the device characteristics that can offer the biggest benefit to improve overall sensitivity.

In Figure 5, the dependence of the receiver sensitivity is graphically plotted for the four important detector characteristics of quantum efficiency, ionization factor, bulk dark current and surface dark current. The overall improvement benefit to the sensitivity is summarized in the bar chart of Figure 6 for ‘nighttime’ operations (low background signal). Here it is seen that the biggest benefit (up to 4 dB) would come from an increase in the quantum efficiency of the front-end detector. The second biggest percentage gain would come from improving the ionization factor. This says that it would benefit from

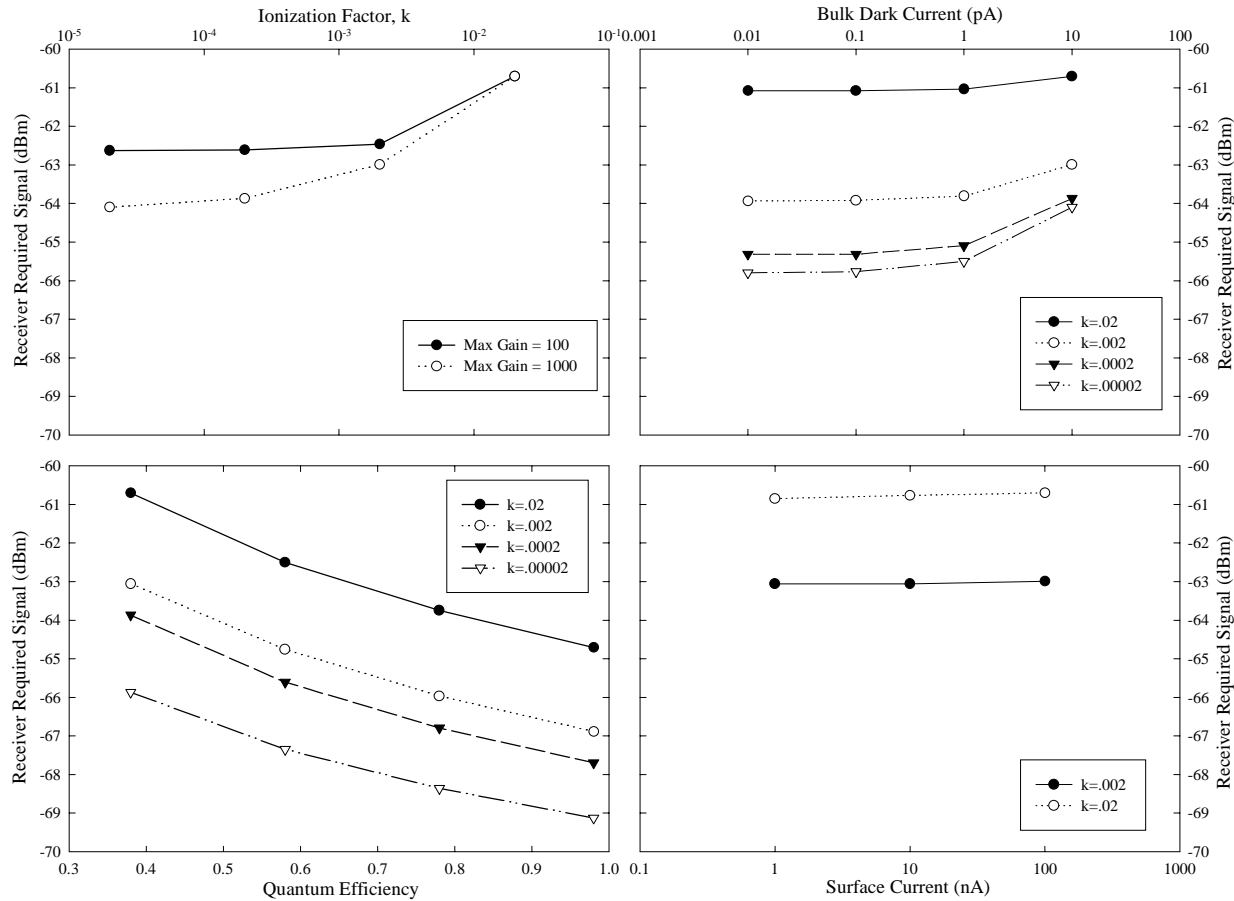


Figure 5. Receiver sensitivity dependence on quantum efficiency, ionization factor, bulk dark current and surface current, at low background signal levels.

amplifying the signal photoelectrons while not multiplying the noise. Reducing the ionization factor by an order of magnitude would result in over 2-dB improvement in the receiver sensitivity. The bulk dark current could yield a 1.5-dB improvement if it is reduced from 10 to 1 pico-Ampere, but, little is gained in further reductions of this noise source.

As the background signal is increased (eg. ‘daytime’ operations), the improvement on the receiver sensitivity from an improved detector is diminished due to the quantum noise limit. In Figure 7 this result is illustrated using three devices, 1) the state-of-the-art high QE Silicon APD (QE = .38,  $k$  = .02), 2) a ‘low- $k$ ’ APD (QE = .38,  $k$  = .007) and 3) an ‘ideal’ APD (QE = .98,  $k$  = .00002). As mentioned in the previous paragraph, for low background signals, the ‘low- $k$ ’ device shows over 2-dB improvement in the receiver sensitivity as compared to the state-of-art device. However, as the background signal increases that improvement is significantly reduced. For example, with a 10-meter diameter telescope under daytime operations, the receiver required signal level would reduce from 440 to 400 photons/pulse to make the same link. This represents only a 0.4-dB improvement in the receiver sensitivity for a 2-dB improvement in device characteristics. This same effect is also seen for an ‘ideal’ APD compared to the state-of-the-art Silicon APD. A potential 10 dB improvement in sensitivity due to improvement in device characteristics, at very low background signal levels, is reduced to an improvement of about 3-dB at high background signal levels due to the quantum noise limit.

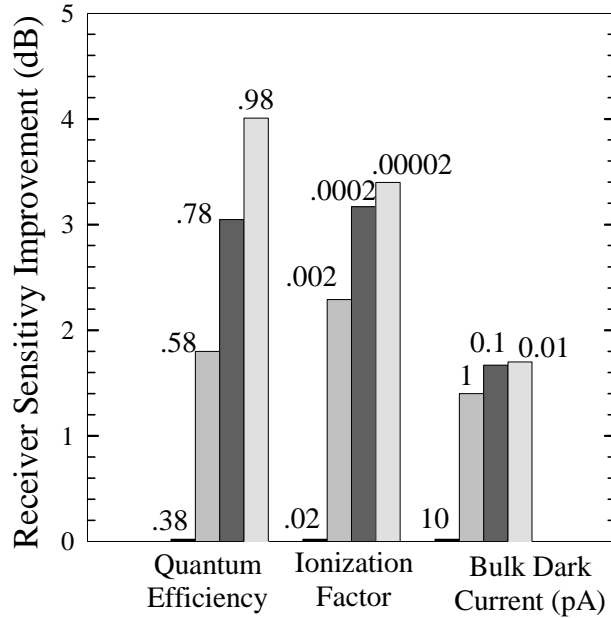


Figure 6. Receiver sensitivity improvement from APD device parameters.

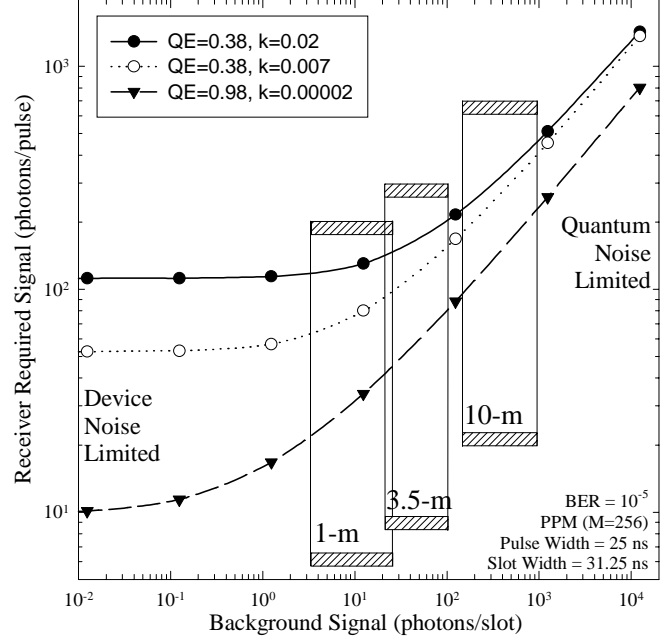


Figure 7. Receiver sensitivity dependence on detector characteristics for three APD devices.

### 2.3.4 Link design/Sensitivity requirement

In this section the link design will be summarized and the opto-electronic receiver sensitivity necessary to make the communications link will be obtained. In order to determine the required receiver sensitivity it is necessary to determine the received signal. This is done by the use of a power link budget, see Table 7.

Link Range	2.99E+08 km	2.00 AU		
Data rate	3.00E+01 kbps	PPM (M = 256)		
Coded BER	1.00E-05	Reed-Solomon Coding		
Transmit power	1.00 W average	9.37 kW (peak)	69.72 dBm	
Transmit losses	72.2 % transmission		-1.41 dB	
Transmitter gain	18.9 urad beamwidth		107.24 dB	
Pointing losses			-2.01 dB	
Space loss			-370.96 dB	
Atmospheric losses	67.2 % transmission		-1.72 dB	
Receiver gain	10.00 m aperture diameter		149.23 dB	
Receiver optics losses	47.4 % transmission		-3.24 dB	
Received signal	6.46E+02 photons/pulse	4.83 nW (peak)	-53.16 dBm	
Required link margin	3.23E+02 photons/slot	pW	3.00 dB	
Required sensitivity	3.23E+02 photons/pulse	2.42 nW (peak)	-56.16 dBm	

Table 7. Typical power link budget for Mars-Earth optical communication link with 10-meter telescope receiver aperture.



In the power link budget all aspects of the link that affect the transmission of signal power are tracked in order to arrive at the detector received signal level. As stated in section 1, the link data rate objective is 30 kbps. Calculating the signal received with a 10-meter telescope and using all stated link characteristics a received signal level of 646 photons/pulse is obtained. Subtracting the required 3-dB link margin, this leads to an opto-electronic receiver sensitivity requirement of 323 photons/pulse.

To meet the required sensitivity for this design the ‘state-of-the-art’ Silicon APD is used as its front-end detector, as long as the background signal is below 400 photons/slot, see Figure 8. This is possible under ‘nighttime’ operations and in ‘daytime’ when the sun is not close to the field-of-view of the telescope, thereby, keeping the sky radiance low. Therefore, a 30 kbps optical link can be made to Mars using the state-of-the-art Silicon APD that is commercially available. If operation is restricted to ‘nighttime’ operations, then the data rate of the link can be increased to 46 kbps for this same detector. Figure 8 also shows that the data rate would be reduced to 20 kbps if the receiver was to experience the highest ‘daytime’ background noise power. Also, if different diameter telescopes were used the data rate would be reduced due to the reduction in received signal power available to the detector.

To improve the receiver performance beyond these values improvements in detector characteristics need to be pursued. The analysis of section 2.2.4, showed that the sensitivity would see the most improvement by increasing the quantum efficiency and by reducing the ionization factor. Two approaches have been proposed. One involves using the Silicon based device with the same QE (=0.38) and reducing the ionization factor to a value of 0.007. The second approach involves the use of the naturally higher QE (>0.90) InGaAs material and improving its ionization coefficient to a value <0.02. In Table 8, the detailed characteristics that each of the proposed devices should have are summarized alongside with the state-of-the-art APD.

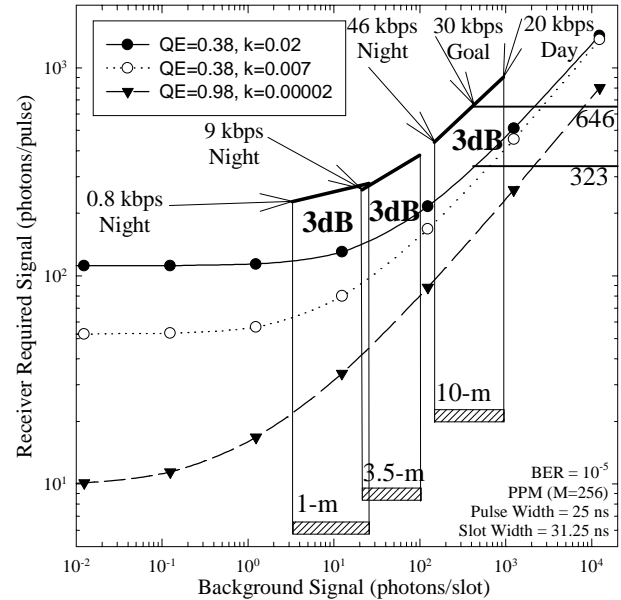


Figure 8. Link receiver sensitivity with 3-dB margin.

Parameter	Si-APD (state-of-art)	Low-k APD (desired)	High-QE APD (desired)
Quantum efficiency (at 1064 nm)	0.38	0.38	0.90
Ionization factor (k-factor)	0.02	0.007	0.02
Bulk dark current, pA	10	0.04	0.04
Surface dark current, nA	100	2.0	2.0
Diameter, mm	1.5	3.0	.03
Capacitance, pF	3	0.25	0.25
Maximum gain	100	100	100
10-90 % rise/fall time, ns	2	0.3	0.3

Table 8. APD detector specifications for opto-electronic receiver design.

This article has shown that the front-end detector dominates the receiver sensitivity. Moreover, in order to further improve the performance of the opto-electronic receiver improvement of the detector characteristics is imperative. The receiver sensitivity improvement realized with the new detectors would translate to an increase in the achievable data rate. For both of these improved APD devices, the increase in the data rate achievable would be the same, see Figure 5. Table 9

summarizes the data rate capabilities of the opto-electronic receiver implemented with the baseline detector and the proposed improved detectors, and a low noise high-impedance pre-amplifier for the three choices of ground telescope diameters being considered.

	Baseline APD	Baseline APD	Improved APD	Improved APD
Telescope Aperture (m)	Data Rate 'Daytime' (kbps)	Data Rate 'Nighttime' (kbps)	Data Rate 'Daytime' (kbps)	Data Rate 'Nighttime' (kbps)
1	0.6	0.8	1.0	1.4
3.5	6	9	9.4	10.9
10	20	46	22.2	50.6

Table 9. Data rate capability of designed opto-electronic receiver for the baseline device of this design and for the improved APD. These values include a 3-dB link margin.

### 3. SUMMARY

The opto-electronic receiver necessary to meet the demands of high capacity deep space optical communication missions has been designed for a Mars-Earth reference mission. The communication link baselined was for a Nd:YAG Q-switched 1 W laser transmitter with PPM modulation ( $M=256$ ). Based on a near-IR-enhanced Silicon APD an opto-electronic receiver was designed that achieved 30 kbps data rates with a BER of  $10^{-5}$ . The receiver sensitivity's dependence on the background noise power and on the APD detector's characteristics was studied and analyzed. The improvement in opto-electronic receiver sensitivity was quantified for improvements in APD quantum efficiency, ionization factor, and bulk dark current. It was shown that novel APD structures that are currently proposed would produce a 3-dB improvement in the receiver sensitivity. It was also shown that as the background signal increased the improvement on the receiver sensitivity from an improved detector is diminished due to the quantum noise limit. Improvements to the APD detector were also studied to describe a design that would achieve over 50 kbps data rates for a Mars-Earth communication link.

### ACKNOWLEDGMENTS

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### APPENDIX A: BLUR CIRCLE

In this Appendix the amount of background light that would be collected during typical daytime operations by an optical telescope on the earth's surface supporting an optical downlink from a Mars orbiter when Mars is at favorable opposition is estimated.

The amount of Marslight or diffuse skylight reaching the detector can be reduced by methods such as restricting the detector's field-of-view (FOV), spectral filtering, and polarization filtering. It is assumed that spectral filtering is applied and that the spectral filter has a full-width pass-band  $\Delta\lambda$  equal to 0.2 nm centered at the downlink wavelength  $\lambda = 1.064 \mu\text{m}$  and completely rejects all wavelengths outside this band. Though polarization filtering can be expected to reduce diffuse skylight by up to 90%<sup>8</sup> during quadrature, diffuse skylight polarization<sup>9</sup> depends in a complicated fashion on solar zenith, observation zenith and azimuth angle, atmospheric aerosol content, and general weather conditions. For this reason, polarization filtering is not treated in this Appendix but is left as a matter of future work. Furthermore, it should be noted that unlike diffuse skylight, Marslight cannot be reduced substantially by polarization filtering because its maximum<sup>10</sup> linear polarization is only about 4%. This Appendix, then, addresses only minimizing background light by restricting the detector's FOV.

While reducing the detector's FOV will always reduce background light, it will also reduce the amount of signal collected. To estimate by how much the detector's FOV may reduce the collected signal for large telescopes, it is necessary to account for the influence of atmospheric turbulence on focus spot size.

Consider a telescope having a circular entrance pupil of diameter  $D$ , a focal length  $f$ , and a detector diameter  $2a$ . The fraction  $P(r)$  of the incident energy collected within the normalized detector radius  $r = a/\lambda F$  is <sup>11</sup>

$$P(r) = 2\pi r \int_0^1 \frac{2}{\pi} \left[ \arccos(v) - v\sqrt{1-v^2} \right] e^{-3.44 \left( \frac{D}{r_0} v \right)^{5/3}} J_1(2\pi r v) dv \quad (1)$$

where  $J_1$  is the Bessel function,  $F = f/D$  is the system focal ratio, and  $r_0$  is the atmospheric correlation length. Obscuration of the telescope primary aperture and scattering from its surface roughness are neglected and the telescope is considered to have 100 % throughput. Formula (1) gives the encircled energy over a long time-average and under the Kolmogorov turbulence approximation; a particular instantaneous realization of the atmospheric turbulence state will have a larger or lesser fraction of the energy enclosed.

Figure 6 is the fraction of encircled energy at a normalized detector radius of  $r = 100$ . For a fixed telescope diameter its horizontal axis can be interpreted as varying atmospheric conditions, with better conditions (larger  $r_0$ ) to the left and worse conditions (smaller  $r_0$ ) to the right. If a change in  $r_0$  causes  $D/r_0$  to increase from  $D/r_0 = 100$  to  $D/r_0 = 200$ , the detector of normalized diameter  $r = 100$ , which encircles 83 % of the energy for the former case, only encircles 44 % of the energy for the latter case.

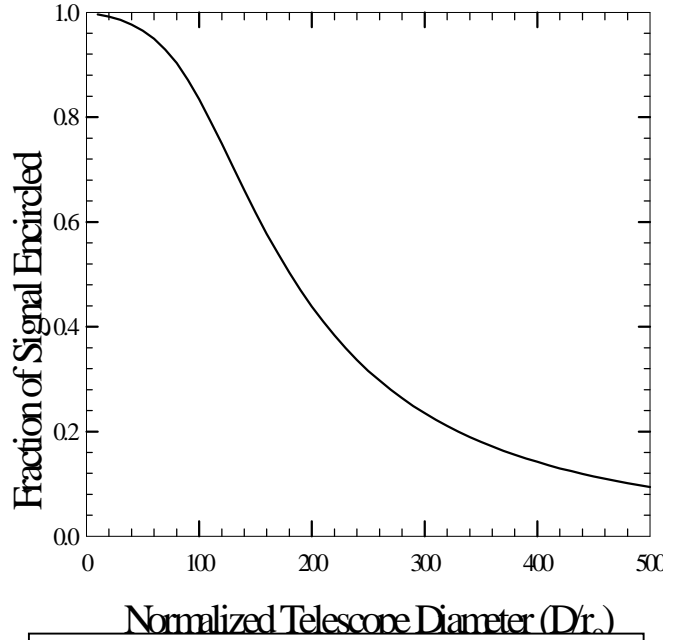


Figure 9. Encircled energy dependence on  $D/r_0$  for a normalized detector radius  $r = 100$ .

## APPENDIX B: DIFFUSE SKY RADIANCE

The typical number of diffuse sky brightness photons reaching the detector, assuming unity throughput, is given by

$$Flux(r) = E \cdot \pi \left( \frac{D}{2} \right)^2 \pi \left( \frac{FOV(r)}{2} \right)^2 \Delta\lambda \frac{1}{hc/\lambda} \Delta T, \quad (2)$$

where  $\Delta T = 1$  ns is the slot duration,  $\delta\lambda = 0.2$  nm is the filter spectral passband, and the FOV is given by

$$FOV(r) = \frac{2a}{f} = \frac{2r\lambda F}{f} = \frac{2r\lambda \frac{f}{D}}{f} = \frac{2r\lambda}{D}. \quad (3)$$

Substituting Eq. 3 into Eq. 2,

$$Flux(r) = E \cdot \left( \frac{\pi}{2} \right)^2 (r\lambda)^2 \Delta\lambda \frac{1}{hc/\lambda} \Delta T, \quad (4)$$

reveals that the flux entering the normalized detector radius is independent of the telescope diameter; although a larger telescope captures more diffuse skylight, the increase is compensated by the reduction in FOV for a fixed focal ratio.

## APPENDIX C: MARSLIGHT

When the field-of-view of the detector is smaller than the angular diameter of Mars, the flux of background photons collected from Marslight is given approximately by the formula

$$Flux_{Mars}(r) = E_{Sun} \cdot \left( \frac{1 \text{ AU}}{R_{MS}} \right) \cdot \Delta\lambda \cdot \pi \left( \frac{FOV}{2} R_{ME} \right)^2 \cdot A \cdot \left( \frac{\pi \left( \frac{D}{2} \right)^2}{2\pi R_{ME}^2} \right)^2 \frac{1}{hc/\lambda} \Delta T \quad (5)$$

where  $E_{Sun} = 0.65 \text{ W/m}^2/\text{nm}$  is the solar illumination at a distance of 1 AU from the sun at  $1.06 \mu\text{m}^{12}$ ,  $R_{MS} = 1.381 \text{ AU}$  is the Mars-Sun distance at Mars perihelion<sup>13</sup>,  $R_{ME} = 5.6 \times 10^{10} \text{ m}$  is the Mars-Earth distance in meters at Mars favorable opposition<sup>13</sup>, and  $A = 0.16$  is the Mars planet-average albedo<sup>12</sup> at  $1.06 \mu\text{m}$ . The model implicit in Eq. 5 for sunlight scattering from Mars at favorable opposition is that scattering occurs uniformly into a hemisphere of space. It is important to note, however, that  $A = 0.16$  is the planet-average Mars albedo and that there is at least a factor of four variability in the Martian albedo with location on Mars<sup>12</sup>. This effect becomes important only when the field-of-view of the detector is comparable to the Mars albedo feature size and is not included in these calculations.

When the field-of-view of the detector is larger than the angular diameter of Mars, Eq. 5 is replaced by

$$Flux_{Mars}(r) = E_{Sun} \cdot \left( \frac{1 \text{ AU}}{R_{MS}} \right) \cdot \Delta\lambda \cdot \pi \left( \frac{\Phi_{Mars}}{2} R_{ME} \right)^2 \cdot A \cdot \left( \frac{\pi \left( \frac{D}{2} \right)^2}{2\pi R_{ME}^2} \right)^2 \frac{1}{hc/\lambda} \Delta T \quad (6)$$

where  $\Phi_{Mars} = 125 \mu\text{rad}$  is the angular diameter of Mars at favorable opposition<sup>13</sup>. In this case the flux of photons received ceases to depend on the detector FOV because the FOV already encompasses Mars.

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